Features

19-3731: Rev 0: 10/92



Low-Noise Precision Operational Amplifiers

General Description

The OP27/OP37 precision operational amplifiers provide lower noise and higher speed with the same input offset and drift specifications as the OP07. Both parts have a $10\mu V$ offset, $0.2\mu V/^{\circ}C$ drift, and 1.8 million gain. Coupled with a low-voltage noise of $3.5nV/\overline{Hz}$ at 10Hz and a low $1/\overline{t}$ noise corner frequency of 2.7Hz, the OP27/OP37 are optimized for accurate amplification of low-level signals. The OP27 features an 8MHz gain-bandwidth product and a $2.8V/\mu s$ slew rate. For applications demanding higher speed, the OP37 has a 63MHz gain-bandwidth product, $17V/\mu s$ slew rate, and is stable at gains of five or more.

An output swing of $\pm 10V$ into 600Ω together with low distortion make the OP27/OP37 ideal for professional audio applications.

For applications requiring greater precision or lower noise than the OP27 or OP37, see the MAX427/MAX437 and the MAX410/MAX412/MAX414 data sheets.

Applications

Low-Noise DC Amplifiers

Microphone Amplifiers

Precision Amplifiers

Tape-Head Preamplifiers

Thermocouple Amplifiers

Low-Level Signal Processing

Medical Instrumentation

Strain-Gauge Amplifiers

High-Accuracy Data Acquisition

♦ 10μV Input Offset Voltage

- ♦ 0.2µV/°C Drift
- ♦ 3nV/√Hz Input Noise Voltage (1kHz)
- ♦ 80nV_{p-p} Noise (0.1Hz to 10Hz)
- ♦ 2.8V/µs Slew Rate (OP27)
- ♦ 17V/µs Slew Rate (OP37)
- ♦ 8MHz Gain-Bandwidth Product (OP27)
- ♦ 63MHz Gain-Bandwidth Product (OP37)

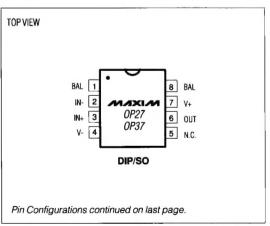
Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE
OP27EP	0°C to +70°C	8 Plastic DIP
OP27FP	0°C to +70°C	8 Plastic DIP
OP27GP	-40°C to +85°C	8 Plastic DIP
OP27GS	-40°C to +85°C	8 SO
OP27EZ	-40°C to +85°C	8 CERDIP
OP27FZ	-40°C to +85°C	8 CERDIP
OP27GZ	-40°C to +85°C	8 CERDIP
OP27EJ	-40°C to +85°C	8 TO-99
OP27FJ	-40°C to +85°C	8 TO-99

Ordering Information continued on last page.

Typical Application Circuit

Pin Configurations



MIXIM

Maxim Integrated Products 1

Call toll free 1-800-998-8800 for free samples or literature.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage
Input Voltage (Note 1) ±22V
Output Short-Circuit Duration
Differential Input Voltage (Note 2)±0.7V
Differential Input Current (Note 2) ±25mA
Continuous Power Dissipation (TA = +70°C)
Plastic DIP (derate 9.09mW/°C above +70°C) 727mW
SO (derate 5.88mW/°C above +70°C) 471mW
CERDIP (derate 8.00mW/°C above +70°C) 640mW
TO-99 (derate 6.67mW/°C above +70°C) 533mW

Operating Temperature Ranges:
OP27/OP37EP/FP 0°C to +70°C
OP27/OP37G_/EZ/EJ/FZ/FJ40°C to +85°C
OP27/OP37A_/B_/C55°C to +125°C
Junction Temperature Range65°C to +150°C
Storage Temperature Range65°C to +150°C
Lead Temperature (soldering, 10 sec) +300°C

Note 1: For supply voltages less than ±22V, the absolute maximum input voltage is equal to the supply voltage.

Note 2: OP27/OP37 Inputs are protected by back-to-back diodes. Current-limiting resistors are not used in order to achieve low noise. If differential input voltage exceeds ±0.7V, the input current should be limited to 25mA.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS $(V_S = \pm 15V, T_A = +25^{\circ}C, unless otherwise noted.)$

PARAMETER	SYMBOL	ODOTA /F ODOTO/F ODO			ODOTA/E ODOTD/E							UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage (Note 3)	Vos			10	25		20	60		30	100	μ٧
Long-Term Vos Stability (Notes 4, 5)	Vos/TIME			0.2	1.0		0.3	1.5		0.4	2.0	μV/Mo
Input Bias Current	IB			±10	±40		±12	±55		±15	±80	nA
Input Offset Current	los			_7	35		9	50		12	75	nA
Input Voltage Range	Ivr		±11.0	±12.3		±11.0	±12.3		±11.0	±12.3		V
Input Resistance - Differential Mode (Note 6)	RIN		1.3	6		0.94	5		0.7	4		МΩ
Input Resistance - Common Mode	RINCM			3			2.5			2		GΩ
Input Noise Voltage (Notes 5, 7)	e _{np-p}	0.1Hz to 10Hz		0.08	0.18		80.0	0.18		0.09	0.25	μ∨р-р
		f ₀ = 10Hz		3.5	5.5		3.5	5.5		3.8	8.0	
Input Noise-Voltage Density (Note 5)	en	f _o = 30Hz		3.1	4.5		3.1	4.5		3.3	5.6	nV/√Hz
		fo = 1kHz		3.0	3.8		3.0	3.8		3.2	4.5	
		f _o = 10Hz		1.7	4.0		1.7	4.0		1.7		
Input Noise-Current Density (Notes 5, 8)	in	f _o = 30Hz		1.0	2.3		1.0	2.3		1.0		pA/√Hz
(110:00 0, 0)		$f_0 = 1kHz$		0.4	0.6		0.4	0.6		0.4	0.6	
		$R_L \ge 2k\Omega$, $V_O = \pm 10V$	1000	1800		1000	1800		700	1500		
Large-Signal Voltage Gain	Avo	$R_L \ge 1k\Omega$, $V_O = \pm 10V$	800	1500		800	1500		400	1500		V/mV
		$R_L \ge 600\Omega$, $V_O = \pm 1V$, $V_S = \pm 4V$ (Note 5)	250	700		250	700		200	500]
Output Voltage Swing	Va	R _L ≥ 2kΩ	±12.0	±13.8		±12.0	±13.8		±11.5	±13.5		
Output Voltage Swing	Vo	R _L ≥ 600Ω	±10.0	±11.5		±10.0	±11.5		±10.0	±11.5		V

ELECTRICAL CHARACTERISTICS (continued) $(V_S = \pm 15V, T_A = +25^{\circ}C, unless otherwise noted.)$

PARAMETER SYMBOL		CONDITIONS	OP27A/E OP37A/E		OP27B/F OP37B/F			0	UNITS			
	i		MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Open-Loop Output Resistance	Ro	V _O = 0, I _O = 0		70			70			70		Ω
Common-Mode Rejection Ratio	CMRR	V _{CM} = ±11V	114	126		106	123		100	120		dB
Power-Supply Rejection Ratio	PSRR	Vs = ±4V to ±18V		1	10		1	10		2	20	μ٧/٧
		f _o = 100kHz, OP27	5.0	8.0		5.0	8.0		5.0	8.0		
Gain-Bandwidth Product (Note 5)	GBP	fo = 10kHz, AvCL≥5, OP37	45	63		45	63		45	63		MHz
(110.00)		fo = 1MHz, AvcL ≥ 5, OP37		40			40			40		
(No. 1) Date (No. 1)	CD	R _L ≥ 2kΩ, OP27	1.7	2.8		1.7	2.8		1.7	2.8	_	1/4
Slew Rate (Note 5)	SR	R _L ≥ 2kΩ, Avc _L ≥ 5, OP37	11	17		11	17		11	17	,	V/μs
Power Dissipation	PD	Vo = 0		90	140		90	140		100	170	mW
Offset Adjustment Range		$R_P = 10k\Omega$		±4.0			±4.0			±4.0		mV

ELECTRICAL CHARACTERISTICS ($V_S = \pm 15V$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		OP27# OP37#			OP27E OP37E			OP270		UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage (Note 3)	Vos			30	60		50	200		70	300	μ٧
Average-Offset Voltage Drift (Note 9)	TCVos			0.2	0.6		0.3	1.3		0.4	1.8	μV/°C
Input Bias Current	lB_			±20	±60		±28	±95		±35	±150	nA
Input Offset Current	los			10	50		14	85		20	135	nΑ
Input Voltage Range	IVR		±10.3	±11.5		±10.3	±11.5		±10.2	±11.5		V
Large-Signal Voltage Gain	AVo	$R_L \ge 2k\Omega$, $V_O = \pm 10V$	600	1200		500	1000		300	800		V/mV
Maximum Output-Voltage Swing	Vo	R _L ≥ 2kΩ	±11.5	±13.5		±11.0	±13.2		±10.5	±13.0		٧
Common-Mode Rejection Ratio	CMRR	V _{CM} = ±10V	108	122		100	119		94	116		dB
Power-Supply Rejection Ratio	PSRR	$V_S = \pm 4.5 V \text{ to } \pm 18 V$		2	16		2	20		4	51	μ٧/٧

ELECTRICAL CHARACTERISTICS (continued)

(Vs = ±15V, TA = TMIN to TMAX, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		OP27E			OP27F OP37F			OP270		UNITS
			MIN	TYP	MAX	MIN	TYP	MAX	MIN	TYP	MAX	
Input Offset Voltage (Note 3)	Vos			20	50		40	140		55	220	μV
Average Offset-Voltage Drift (Note 9)	TCVos			0.2	0.6		0.3	1.3		0.4	1.8	μV/°C
Input Bias Current	lB			±14	±60		±18	±95		±25	±150	nA
Input Offset Current	los			10	50		14	85		20	135	nA
Input Voltage Range	IVR		±10.5	±11.8		±10.5	±11.8		±10.5	±11.8		V
Large-Signal Voltage Gain	Avo	$R_L \ge 2k\Omega$, $V_0 = \pm 10V$	750	1500		700	1300		450	1000		V/mV
Output Voltage Swing	Vo	R _L ≥ 2kΩ	±11.7	±13.6		±11.4	±13.5		±11.0	±13.3		٧
Common-Mode Rejection Ratio	CMRR	V _{CM} = ±10V	110	124		102	121		96	118		dB
Power-Supply Rejection Ratio	PSRR	$V_S = \pm 4.5 V \text{ to } \pm 18 V$		2	15		2	16		2	32	μ٧/٧

Note 3: Vos is measured approximately 0.5 seconds after application of power.

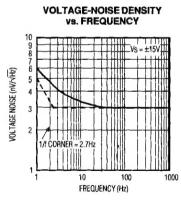
Note 4: Long-term input offset voltage stability refers to the average trend line of Vos vs. Time over extended periods after the first 30 days of operation.

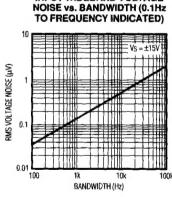
Guaranteed by design.

Note 5: Guaranteed by design.
 Note 6: Guaranteed by input bias current.
 Note 7: See test circuit and frequency response curve for 0.1Hz to 10Hz tester (Figures 1, 6).
 Note 8: See test circuit for current-noise measurement (Figure 2).
 Note 9: The TCVos performance is within the specifications unnulled or when nulled with Rp = 8kΩ to 20kΩ. TCVos is sample tested to 0.1% AQL for A/E grades. B/C/F/G are guaranteed by design.

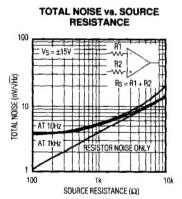
Typical Operating Characteristics

(TA = +25°C, unless otherwise noted)

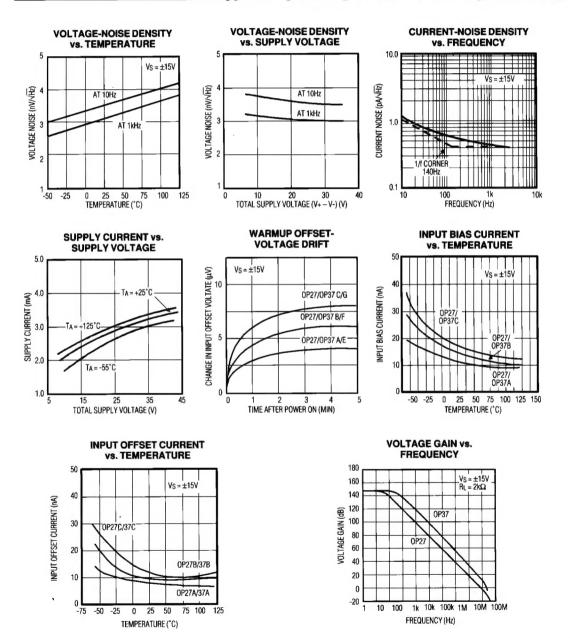


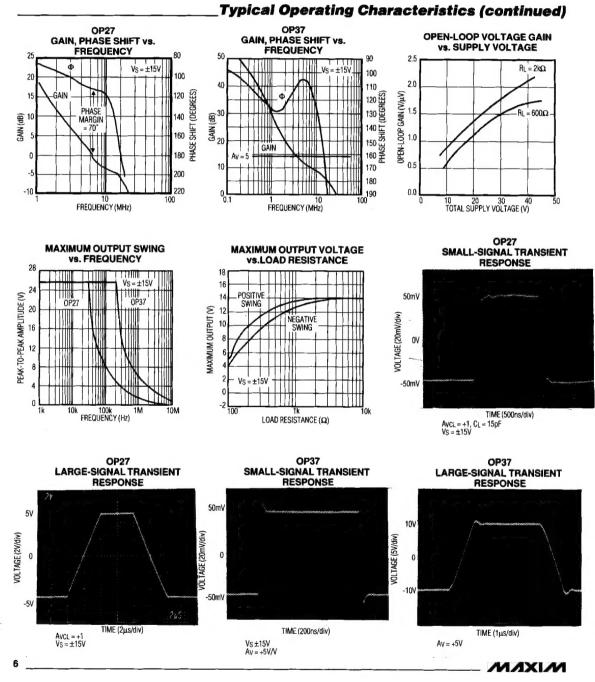


INPUT WIDEBAND VOLTAGE

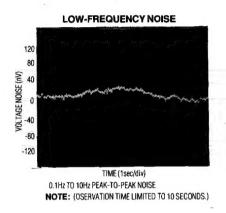


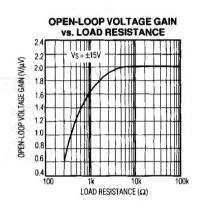
Typical Operating Characteristics (continued)





Typical Operating Characteristics (continued)





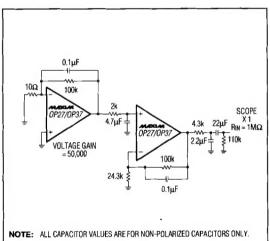


Figure 1. Voltage-Noise Test Circuit (0.1Hz to 10Hz)

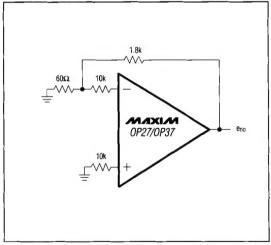


Figure 2. Current-Noise Test Circuit

Applications Information

The OP27/OP37 provide stable operation with load capacitances of up to 2nF and $\pm 10\text{V}$ output swings; larger capacitances should be decoupled with a 50Ω series resistor inside the feedback loop. The OP27 is unity-gain stable and the OP37 is stable at gains of five or greater.

Thermoelectric voltages generated by dissimilar metals at the input terminals degrade the drift performance. Connections to both inputs should be maintained at the same temperature for best operation.

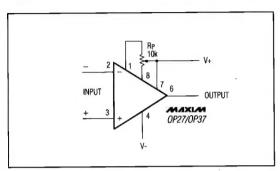


Figure 3. Offset Nulling Circuit

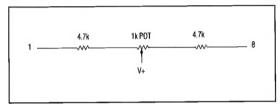


Figure 4. Alternate Offset-Voltage Adjustment

Offset-Voltage Adjustment

Input offset voltage (Vos) is trimmed at the wafer level. If Vos adjustment is necessary, a $10 k\Omega$ trim potentiometer (pot) may be used and will not degrade TCVos (Figure 3). Other trim pot values from $1 k\Omega$ to $1 M\Omega$ can be used with a slight degradation $(0.1 \mu V)^c$ C to $0.2 \mu V)^c$ C) of TCVos. Adjusting, but not zeroing, Vos creates a drift of approximately (Vos/300) $\mu V)^c$ C. For example, the change in TCVos is $0.33 \mu V)^c$ C if Vos is adjusted to $100 \mu V$. The adjustment range with a $10 k\Omega$ trim pot is $\pm 4 mV$. For a smaller range, reduce nulling sensitivity by connecting a smaller pot in series with fixed resistors; for example, Figure 4 has a $\pm 280 \mu V$ adjustment range.

Noise Measurements

To measure the $80nV_{p-p}$ noise specification of the OP27/OP37 in the 0.1Hz to 10Hz range, observe the following precautions:

- 1. The device must warm up for at least five minutes. Figure 5 shows how Vos typically increases $4\mu V$ with increases in chip temperature after power-up. In the 10sec measurement interval, temperature-induced effects can exceed 10nV.
- 2. For similar reasons, the device must be well-shielded from air currents, including those caused by motion. This minimizes thermocouple effects.

- 3. As shown in Figure 6, the 0.1Hz corner is defined by only one zero. A maximum test time of 10sec acts as an additional zero to eliminate noise contributions from the frequency band below 0.1Hz.
- 4. A noise-voltage-density test is recommended when measuring noise on a large number of units. A 10Hz noise-voltage-density measurement correlates well with a 0.1Hz to 10Hz peak-to-peak noise reading, since both results are determined by the white noise and the location of the 1/f corner frequency.

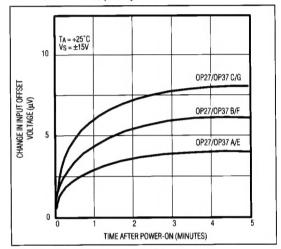


Figure 5. Warm-Up Offset Voltage Drift

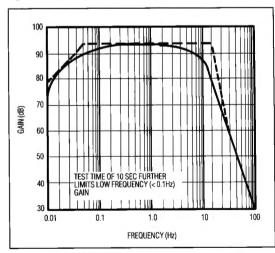


Figure 6. 0.1Hz to 10Hz V_{p-p} Noise Tester Frequency Response

Unity-Gain Buffer Applications (OP27 Only)

Figure 7 shows the circuit and output waveform with R1 \leq 100 Ω , and the input driven with a fast, large signal pulse (>1V).

During the fast rise portion of the output, the input protection diodes short the output to the input, and a current, limited only by the output short-circuit protection, is drawn by the signal generator. With $R_f \ge 500\Omega$, the output is capable of handling the current required ($I_L \le 20$ mA at 10V) and a smooth transition occurs.

When $R_f \ge 2k\Omega$, a pole created with R_f and the amplifier's input capacitance (8pF) causes additional phase shift and reduces phase margin. A small capacitor (20pF to 50pF) in parallel with R_f eliminates this problem.

Comments on Noise

The OP27/OP37 are very low-noise amplifiers. They have outstanding input voltage noise characteristics by operating the input stage at a high quiescent current. Input bias and offset currents, which would normally increase with the quiescent current, are minimized by bias-current cancellation circuitry. The OP27/OP37A and E grade devices have IB and IOs of only ±40nA and 35nA respectively at +25°C. This is particularly important with high source-resistances.

Voltage noise is inversely proportional to the square-root of bias current, but current noise is proportional to the square-root of bias current. The OP27/OP37 low-noise advantages are reduced when high source resistors are used.

Total noise = [(voltage noise)² + (current noise x Rs)² + (resistor noise)²]^{1/2}

Figure 8 shows noise vs. source resistance at 1kHz. To use this plot for wideband noise, multiply the vertical scale by the square-root of the bandwidth. The OP27/OP37 maintains low input noise voltage with Rs < 1k Ω . With Rs > 1k Ω , total noise increases and is dominated by the resistor noise, not the current or the voltage noise. It is only with Rs \geq 20k Ω that current noise dominates. Current noise is not important for applications with Rs < 20k Ω . The OP27/OP37 has lower total noise than the MAX400/OP07 for Rs < 10k Ω . As Rs increases, the crossover between the OP27/OP37 and the MAX400/OP07 noise occurs in the Rs = 15k Ω to 40k Ω region.

Figure 9 shows 0.1Hz to 10Hz peak-to-peak noise. Here, resistor noise is negligible and current noise (in) becomes important, because in $\simeq 1/\sqrt{f}$. The crossover with the MAX400/OP07 occurs in the Rs = 3k Ω to 5k Ω range,

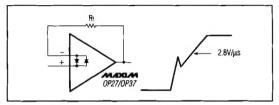


Figure 7. Pulsed Operation of Unity-Gain Buffer

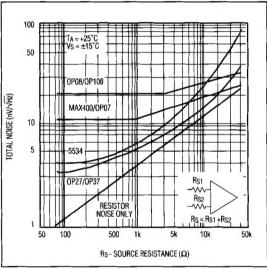


Figure 8. Noise vs. Source Resistance (Including Resistor Noise) at 1kHz

depending on whether balanced or unbalanced source resistors are used (at $3k\Omega$ the IB and IQS error can be three times the VQS specification). For low-frequency applications, the MAX400/OP07 is better than the OP27/OP37 when Rs > $3k\Omega$, except when gain error is important. Figure 10 illustrates the 10Hz noise. As expected, the results fall between those of the previous two figures.

For reference, typical source resistances of some signal sources are listed in Table 1.

Table 1. Signal Source vs. Source Impedance

DEVICE	SOURCE IMPEDANCE	COMMENTS
Strain Gauge	<500Ω	Typically used in low-frequency applications.
Magnetic Tapehead	< 1500Ω	Low IB is very important to reduce self-magnetization problems when direct coupling is used. OP27 IB can be neglected.
Linear Variable Differential Transformer	< 1500Ω	Used in rugged servo-feed- back applications. Bandwidth of interest is 400Hz to 5kHz.

Table 2. Open-Loop Gain vs. Frequency

OPEN-LOOP GAIN							
FREQUENCY AT:	OP07	OP27	OP37				
3Hz	100dB	124dB	125dB				
10Hz	100dB	120dB	125dB				
30Hz	90dB	110dB	124dB				

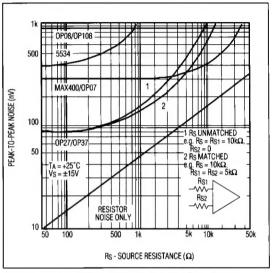


Figure 9. Peak-to-Peak Noise (0.1 to 10Hz) vs. Source Resistance (Includes Resistor Noise)

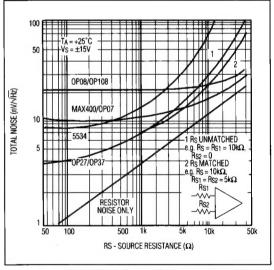


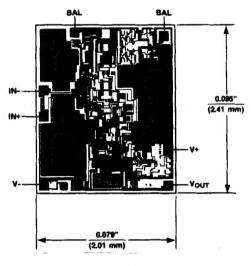
Figure 10. 10Hz Noise vs. Source Resistance (Includes Resistor Noise)

Ordering Information (continued)

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PART	TEMP. RANGE	PIN-PACKAGE
OP27GJ	-40°C to +85°C	8 TO-99
OP27AZ	-55°C to +125°C	8 CERDIP*
OP27BZ	-55°C to +125°C	8 CERDIP*
OP27CZ	-55°C to +125°C	8 CERDIP*
OP27AJ	-55°C to +125°C	8 TO-99*
OP27BJ	-55°C to +125°C	8 TO-99*
OP27CJ	-55°C to +125°C	8 TO-99*
OP37EP	0°C to +70°C	8 Plastic DIP
OP37FP	0°C to +70°C	8 Plastic DIP
OP37GP	-40°C to +85°C	8 Plastic DIP
OP37GS	-40°C to +85°C	8 SO
OP37EZ	-40°C to +85°C	8 CERDIP
OP37FZ	-40°C to +85°C	8 CERDIP
OP37GZ	-40°C to +85°C	8 CERDIP
OP37EJ	-40°C to +85°C	8 TO-99
OP37FJ	-40°C to +85°C	8 TO-99
OP37GJ	-40°C to +85°C	8 TO-99
OP37AZ	-55°C to +125°C	8 CERDIP*
OP37BZ	-55°C to +125°C	8 CERDIP*
OP37CZ	-55°C to +125°C	8 CERDIP*
OP37AJ	-55°C to +125°C	8 TO-99*
OP37BJ	-55°C to +125°C	8 TO-99*
OP37CJ	-55°C to +125°C	8 TO-99*

^{*}Contact factory for availability and processing to MIL-STD-883.

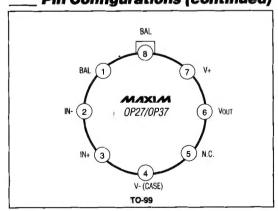
Chip Topography



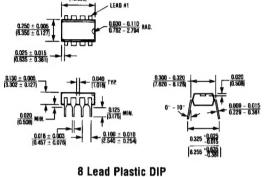
OP27/OP37

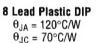
SUBSTRATE CONNECTED TO V-

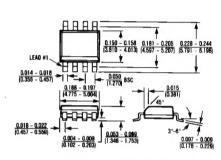
Pin Configurations (continued)



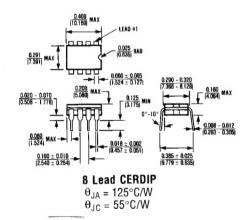
Package Information

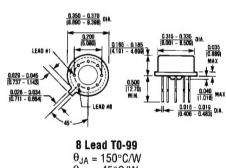






8 Lead Small Outline $\theta_{JA} = 170^{\circ}C/W$ $\theta_{JC} = 80^{\circ}C/W$





 $\theta_{JC} = 45^{\circ}C/W$

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